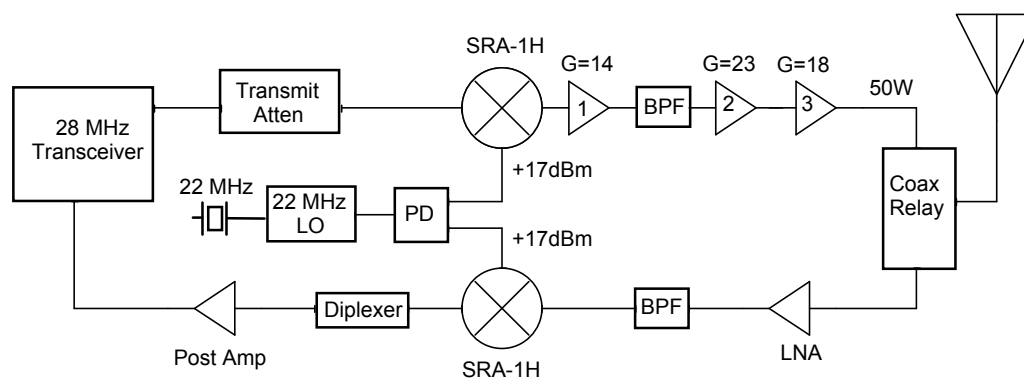
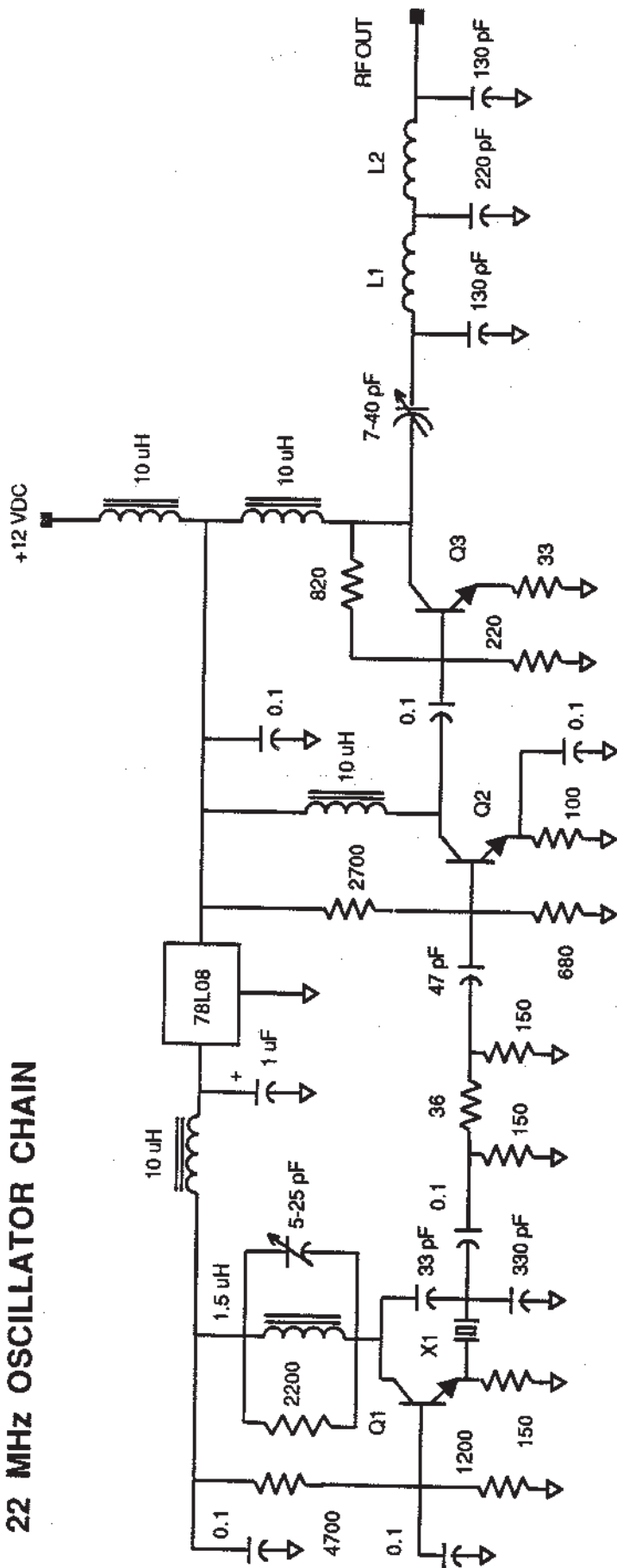


# 50 MHz Transverter WB3JYO design built by WA3JUF 1995



WB3JYO design except:  
Amp1 = SD1520  
LNA = MGF1302  
LO = w/ 60 degC PTC thermistor  
and SD1520 in output  
PD = Mini-Circuits ZFSC-2  
BPF=YU1LM (2007)

# 22 MHz OSCILLATOR CHAIN



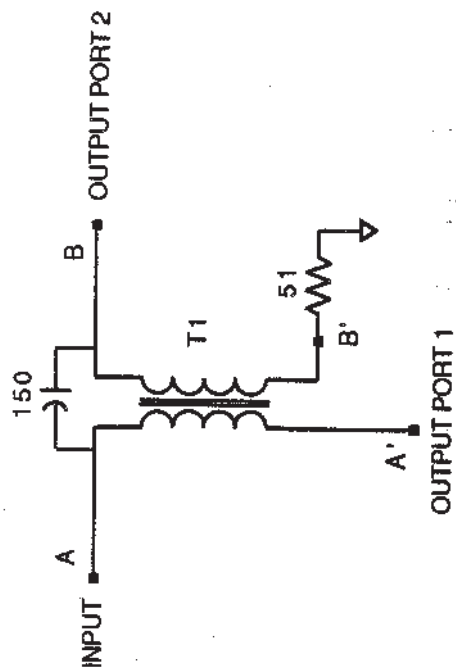
- Q1, Q2 2N5179
- Q3 2N3866 (HEATSINK REQUIRED)
- X1 22 MHz SERIES RESONANT
- L1, L2 12 t #22 ENAM., 1/4 INCH I.D.

RESISTORS IN OHMS  
CAPACITORS IN uF

22 MHZ OSCILLATOR CHAIN  
Pout > +20 dBm  
HARMONICS > 40 dB DOWN

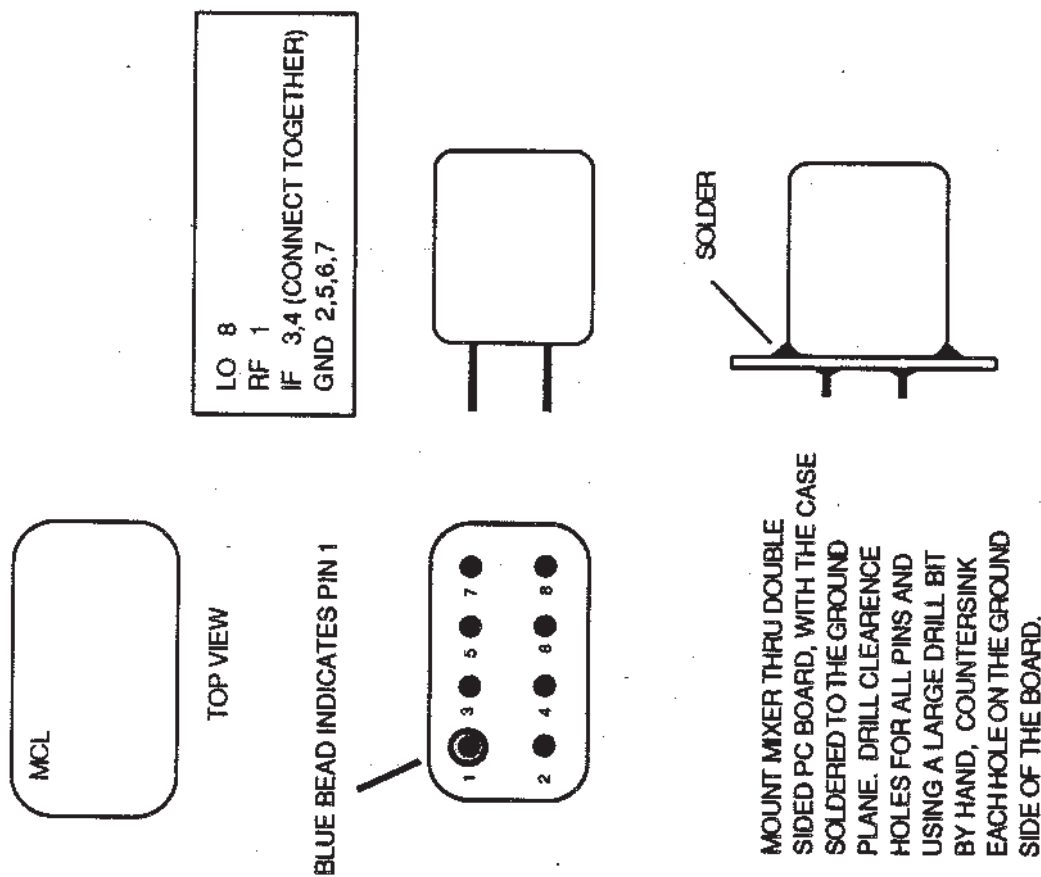
WB3JYO 10-90

FIG. 1 LOW LOSS POWER SPLITTER



T1 IS MADE FROM 7 TURNS #28 ENAMEL WIRE WOUND BIFILAR ON AN AMIDON T37-2 TOROID. THREAD TWO PIECES OF WIRE THRU THE TOROID AND, WHILE HOLDING THESE PARALLEL, WIND 7 TURNS. USE AN X-ACTO KNIFE TO STRIP THE INSULATION FROM EACH END. LABEL THE WIRES ON ONE END AND THEN USE AN OHM METER TO IDENTIFY THE OPPOSITE ENDS. CONNECT THE SPLITTER AS SHOWN.

FIG. 2 MINI-CIRCUITS SRA-1H

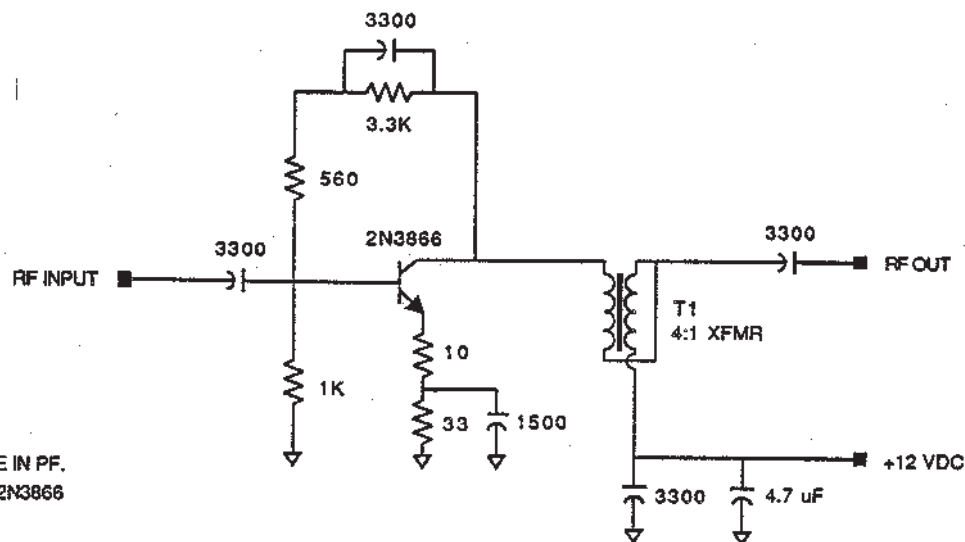


## A 50 MHz 2N3866 LINEAR AMPLIFIER

A 2N3866 amplifier stage follows the mixer in the 50 MHz transmit converter. An inexpensive 2N3866 was chosen since it is a commonly available device having excellent linearity; when driven to +7 dBm in the 50 MHz transmit chain all intermod products are more than 60 dB down. The amplifier has 15 dB gain and yet is stable into a 10:1 mismatch.

As with all of the stages, a point-to-point "dead bug" construction technique is recommended. Care must be taken to keep all lead lengths to a reasonable minimum. All resistors are 1/4 watt. Remember to keep the 2N3866 away from ground since the case is the collector. Use a small push-on type heatsink fashioned from tin or copper as shown. The transformer is wound as shown in the pictorial - wire size is not critical.

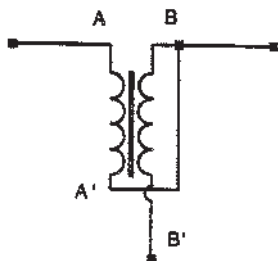
### 50 MHz 2N3866 LINEAR AMPLIFIER



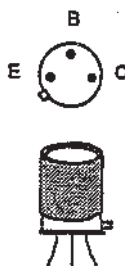
#### NOTE

RESISTANCE IN OHMS, CAPACITANCE IN PF.  
USE PUSH-ON TYPE HEATSINK ON 2N3866

T1 8t #28 BIFILAR ON T30-6 TOROID.  
CONNECT TRANSFORMER AS SHOWN.



#### 2N3866



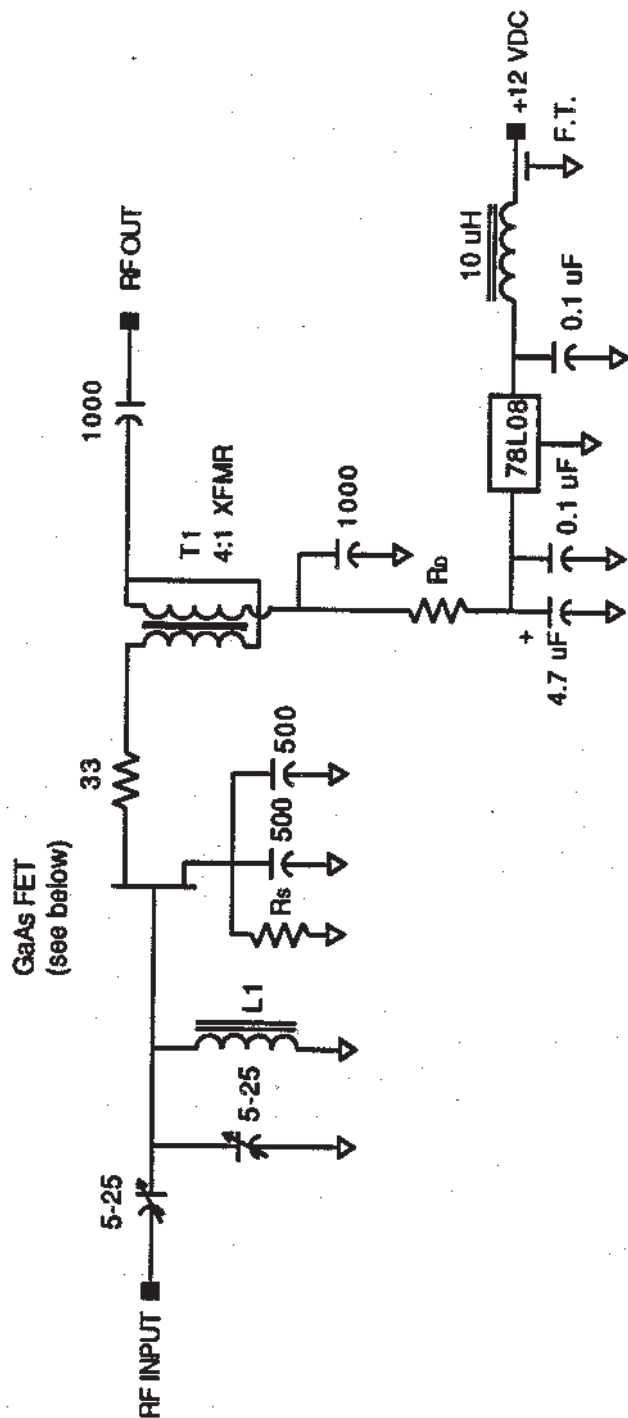
HEATSINK MADE  
FROM COPPER

ICQ = 40 mA  
15 dB GAIN @ 50 MHz  
P1dB = +20 dBm  
STABLE INTO 10:1 VSWR  
IMD < -60 dBc @ +7dBm Pout

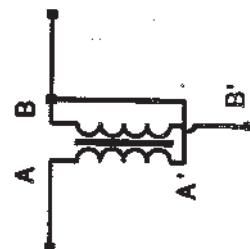
WB3JYO 10-89



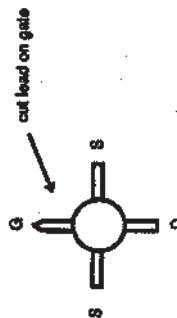
# 50 MHZ LOW DISTORTION PREAMP



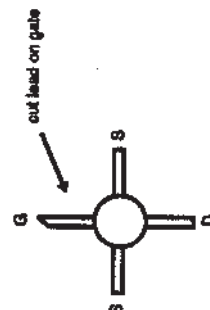
NOTE  
RESISTANCE IN OHMS, CAPACITANCE IN PF  
L1 12 #28 ENAMEL ON T44-10 TOROID  
T1 84 #32 ENAMEL BIFILAR WOUND ON T26-8 TOROID  
CONNECT TRANSFORMER AS SHOWN



ATF13484 TOP VIEW



NE72084 TOP VIEW



**TYPICAL BIAS VALUES**  
Biasing for each transistor has been optimized for lowest distortion in the above circuit. Use the resistor values as a starting point and adjust for the recommended VDS and ID.

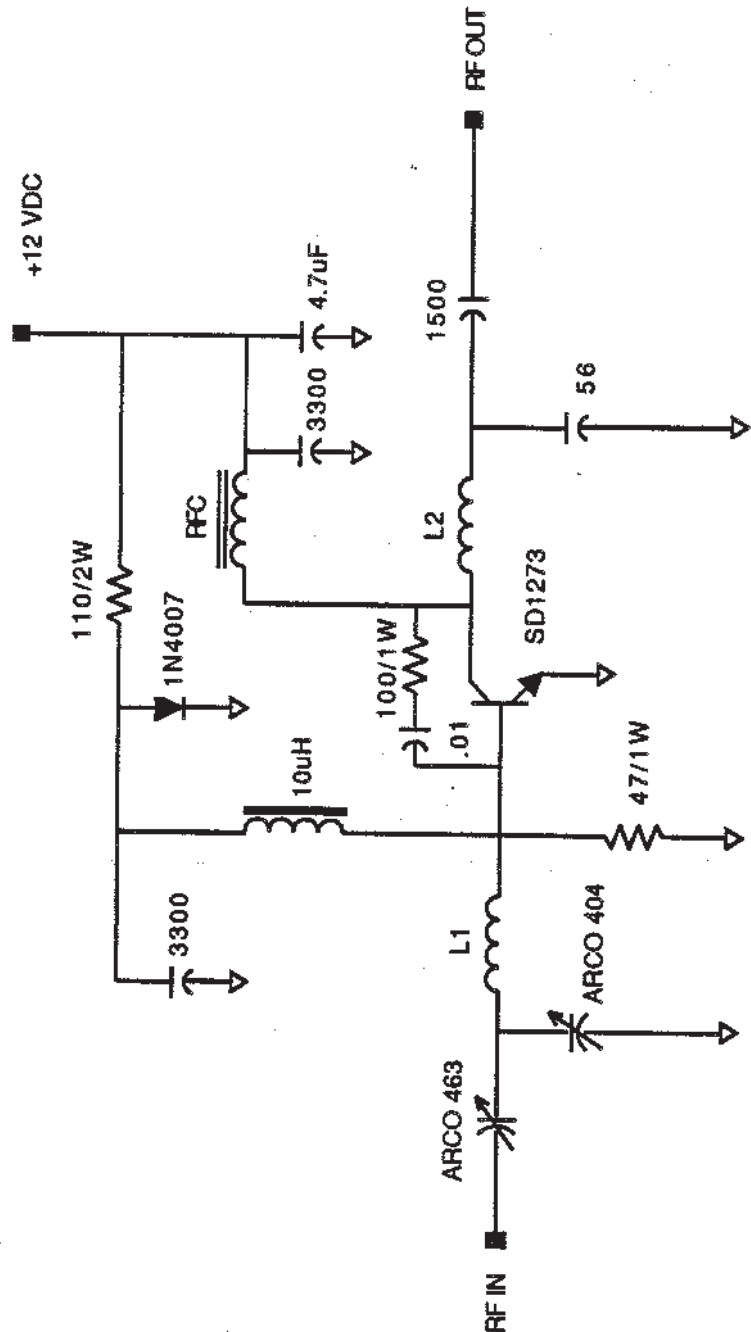
ATF13484 NE72084

RS	47	47
RD	220	39

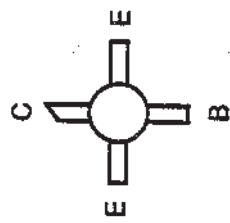
VDS	3.5V	4V
ID	15 mA	40 mA
P1dB	+9 dBm	+14 dBm
P3	+22 dBm	+24 dBm
GAIN	22 dB	26 dB
NF	0.7 dB	0.5 dB

WB3JVO 4-91

# 50 MHz SD1273 LINEAR AMPLIFIER



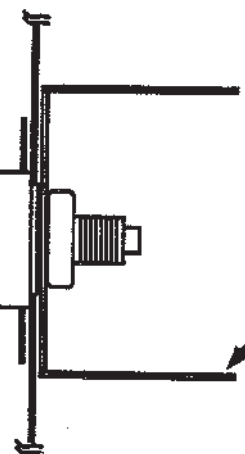
THOMPSON SD1273



## NOTE

- L1 5t #20 AWG, 0.125 ID
  - L2 SAME
  - RFC 10t #26 AWG on 100 ohm/1W
- MOUNT DIODE ON TRANSISTOR CAP AS SHOWN

DRILL CLEARANCE HOLE  
THRU PCB TO ALLOW TRANSISTOR  
BASE TO FIRMLY CONTACT HEATSINK



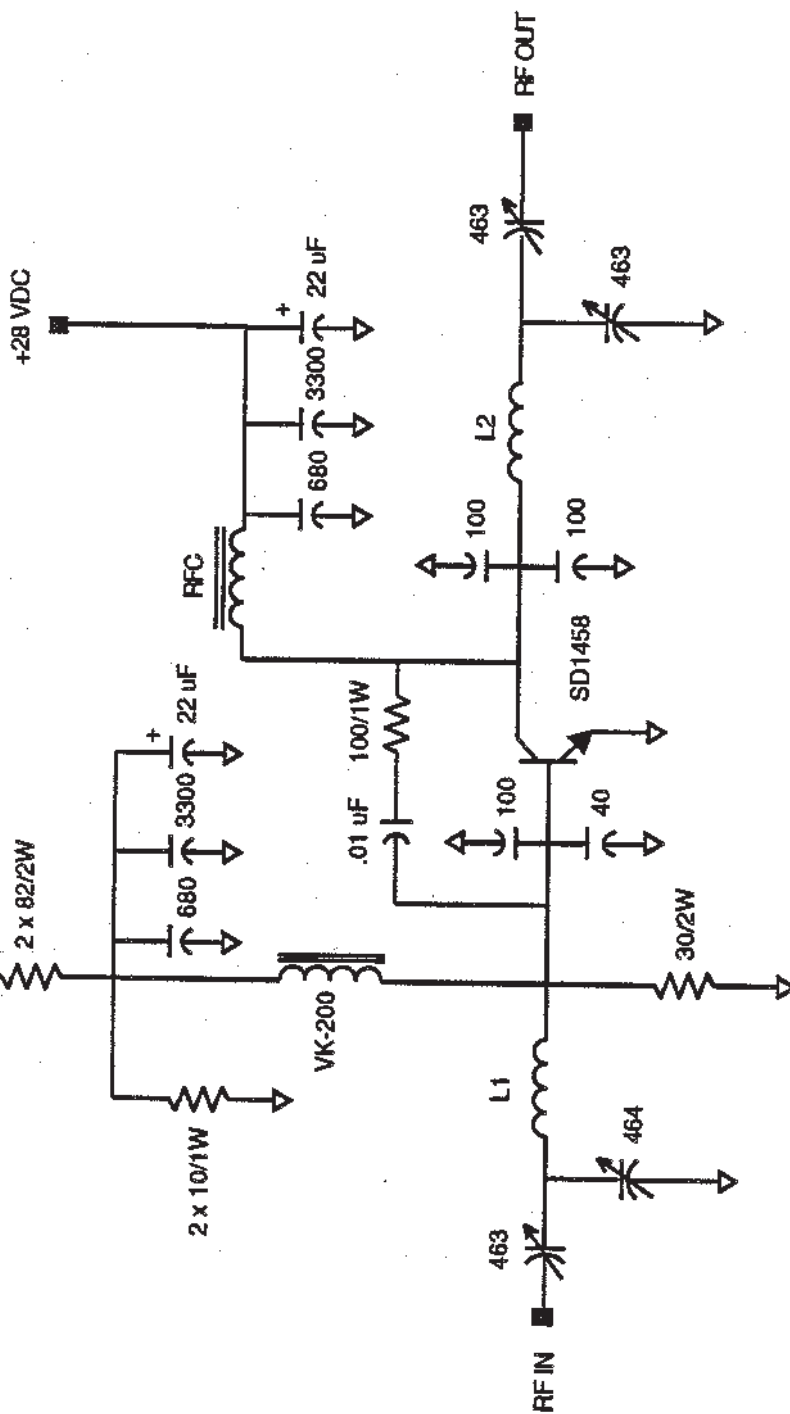
CIRCUIT MAY BE BUILT USING  
POINT-TO-POINT CONSTRUCTION.  
USE WIRE "THRU'S" TO PCB BACK-SIDE  
FOR GOOD GROUNDING. MOUNT DIODE  
ON TRANSISTOR CAP AS SHOWN

GAIN = 23 dB @ 50 MHz  
Icq = 200 mA  
P1dB = +39 dBm  
STABLE INTO 10:1 VSWR  
IMD < -45 dBc @ +23 dBm

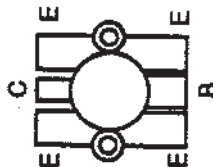
WB3JYO 11-90

# 50 MHz SD1458 LINEAR AMPLIFIER

BIAS (+5 ON XMIT)



THOMPSON SD1458



PAGE 8

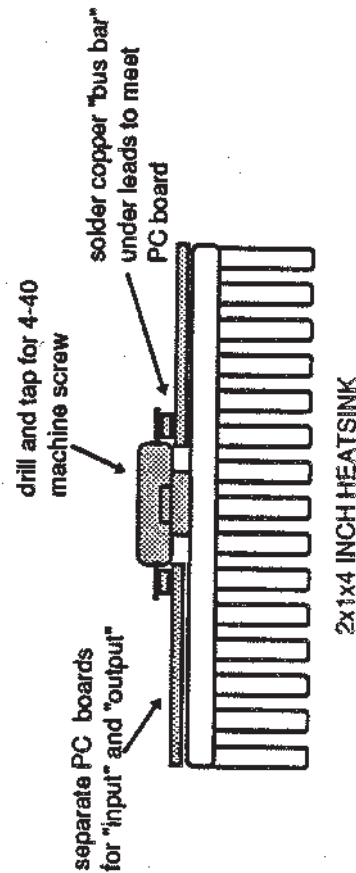
## NOTE

Resistance in ohms, capacitance in pF unless otherwise specified

All variable capacitors are ARCO type (#463 is 20-180 pF, #464 is 45-280 pF)

L1 = 2 turns #20 AWG on 3/8 inch ID  
L2 = 4 turns

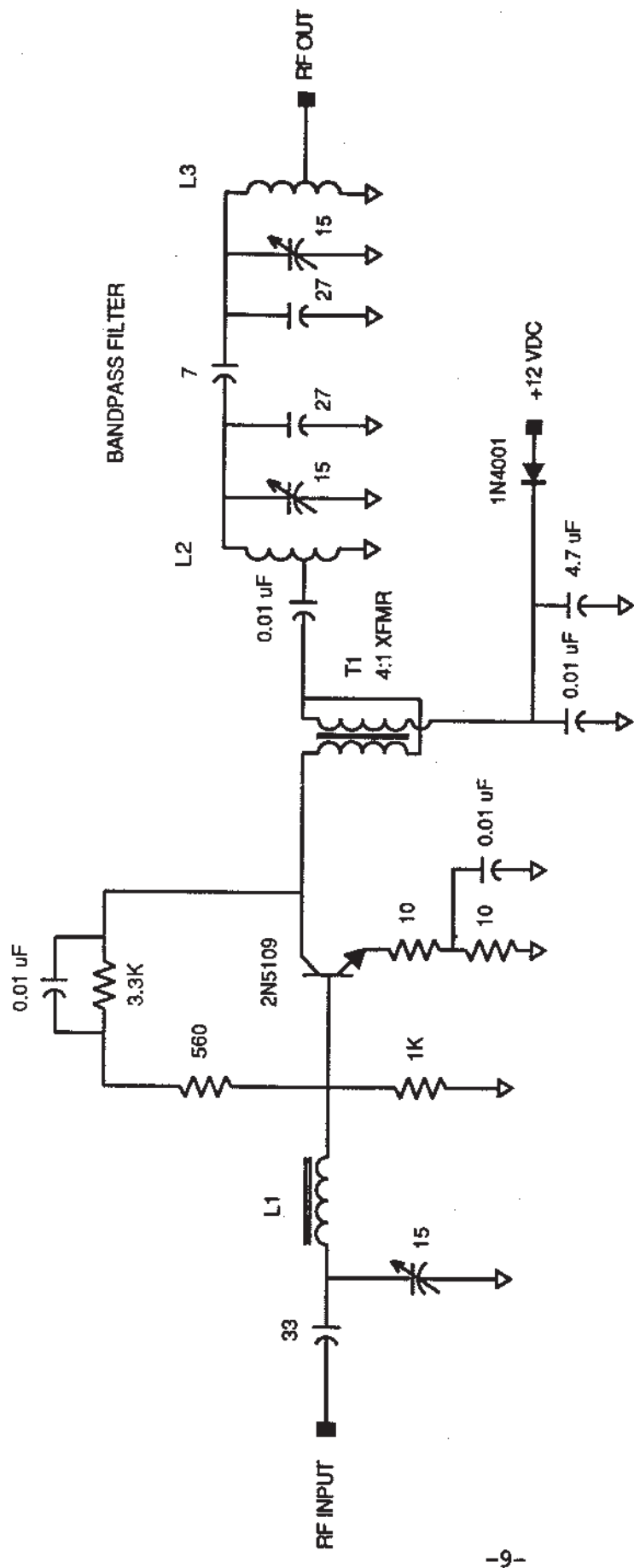
RFC consists of 10 turns #20 AWG enameled wire wound on a 100 ohm/2 W resistor.



GAIN = 18dB @ 50 MHz  
P1dB = +47 dBm  
Icq = 1.5 A  
STABLE INTO 10:1 VSWR  
IMD < -44 dBc @ +40 dBm

WB3JYO 2-91

# 28 MHZ LOW-DISTORTION POST AMPLIFIER



## NOTE

CAPACITANCE IN pF UNLESS NOTED.

USE PUSH-ON TYPE HEATSINK ON 2N5109

L1 15#26 ENAMEL ON T37-6 TOROID

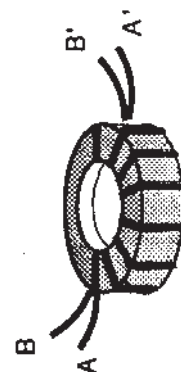
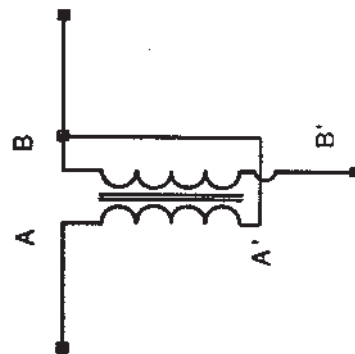
L2, L3 15#26 ENAMEL ON T37-6,

TAP AT 4# FROM GROUND

T1 20#30 BIFILAR ON T37-10 TOROID.

CONNECT TRANSFORMER AS SHOWN.

2N5109



HEATSINK MADE  
FROM COPPER

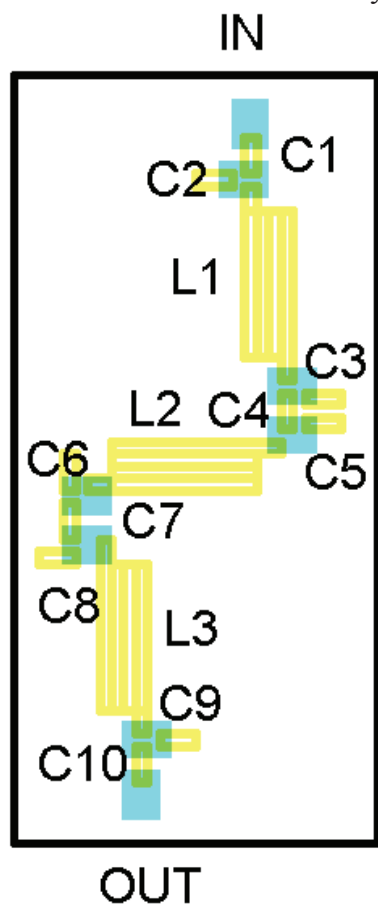
GAIN = 15 dB @ 28 MHz  
I<sub>Q</sub> = 60 mA  
P<sub>1dB</sub> = +23dBm  
Third Order Intercept Point = +32 dBm  
STABLE INTO 10:1 VSWR

WB3JYO 12-91

# *Band Pass Filters for 6m*

**Dipl ing Tasić Siniša –Tasa YU1LM/QRP**

At local HAM forum BalkanDX was discussion about BP (band pass) filters for the 6m few years ago. Here are my proposals band pass filters for 6m band. First filter is with minimum IL (insertion loss) lower than 0.7 dB. This filter is possible to use except in the receiver and also in the transmitter chain. Second filter is with moderate IL loss less than 1.5 dB. The third filter is with max selectivity and IL is less than 2.5 dB. Realization is a very simple and non-critical. Please take care only about physical coils placement. It is necessary to prevent mutual coupling between coils or this coupling has to be minimized. This unwanted coupling will destroy out of band selectivity. Best way for montage is free space construction but it is possible make a PCB. Take care that coils are perpendicular one to each other. See my proposal down please



**PCB DIMENSION 40 X 80 mm**

All coils are equal and wound with isolated copper (Cu) wire diameter 1.2 mm (AVG18) without support on body diameter 14 mm. Coils length is 5 mm with 4 turns. For coils with smaller inductivity than 320 nH it is necessary only to make carefully space between wounds.

1. Minimum IL 6 m BP filter –This filter can be at RX input or TX output. BP Bandwidth is 8.7 MHz (-3 dB)

